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A COMPACT DESIGN OF WI-MAX APPLICATION FOR CPW BASED FED KOCH FRACTAL ANTENNA SLOT

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Abstract

The double large band CPW took care of adjusted Koch fractal space receiving wire proposed in this article is sensible for WLAN and Wi-MAX errands. The running repeat of a three-sided opening radio wire is obscured here using the Koch new conveyance system, achieving flexible receiving wire tests the impedance and radiation instances of the proposed radio wire, showing that an invigorated Koch fractal space receiving wire has an impedance information transmission of 2.38 to 3.95 5 GHz and 4.95 5-6.05 GHz cautious 2.4/5.2/5.8 GHz WLAN gatherings and the 2.5 GHz WLA. In the whole working band, the receiving wire composed radiation incorporation has an increment of more than 2.0 dBi. The disclosures are found and connected with accurate people from their own families.

CPW-dealt with opening radio wires, printed fractal space receiving wires, wide-band receiving wires, and WLAN receiving wires are cases of document terms.

1. INTRODUCTION

The fundamental for low-profile, lightweight, and insignificant cost broadband radio wires has extended actually as short distance far away frameworks relationship, for instance, far away district, has gotten more norm (WLAN). WLANs are planned to work in the 2.4 GHz (2.4-2.48 GHz) and 5 GHz repeat social affairs (5.15-5.35 GHz and 5.725-5.825 GHz in the United States and 5.15-5.35 GHz and 5.47-5.725 GHz in Europe). Wi-MAX (Worldwide Interoperability for Microwave Access) is a quick deployable, inconsequential cost broadband far off structures affiliation standard that works in the 2.5-2.69/3.4-3.69/5.25-5.85 GHz get-togethers. Since these standards may be used in

different relationship at the same time, a singular radio wire that covers the two social affairs is required. A co-planar waveguide (CPW) feed is better sensible for lightweight distant constructions affiliation applications because to its portions, for instance, uni-planar turn of events, fast assembling, and circuit joining. Specific opening evaluations like square shape, rectangular, three-sided, trapezoidal, underhanded, contorted, and others have been seen in literature[2]-[11] in blend in with either a rectangular, fork-like, or round tuning stub, invigorated for wide-band working. Using a multipleresonance-

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production overseeing instrument, information move cutoff may be extended. Then, by changing the opening between the tuning stub and the including field, the impedance progress beginning with one reverberating mode then onto coming up next is restricted, achieving wide band improvement. Since the most lessened resonance of a wide opening receiving wire is obliged by the opening boundary[9]-[11], the space filling pondered the Koch turns utilized in the movement of traditionalist and multi-bandpatch radio wires may be extended[12]. Given the lower working frequencies[6]-[9], the proposed radio wire, expected WLAN/Wi-MAX executions, everything considered beats the ultra wide band opening receiving wire smoothed out for the FCC upheld UWB band (3.1-10.6 GHz) to the degree adaptability. While a wide band radio wire working some spot in the degree of 2.3 and 6 GHz will do what needs to be done, a twofold band receiving wire will by a long shot decrease the req[6]-[9]. For the divert improvement in this letter, a half-rehash tuning opening is gotten along with a Koch wide-band space radio wire. The radio wire performs twofold wide-band headway, satisfying the WLAN and Wi-MAX packs simultaneously while staying aware of a compact profile, on account of Koch fractal-based space plan.

2. PLAN

The proposed fortified Koch space receiving wire's course of action for twofold band movement is tended to in Fig. 1. A 50CPW feed and a tuning stub implanted with a U-outlined cut on a low scene substrate with relative permittivity and thickness feed the radio wire. Koch snowflake opening that has been changed. AnsoftHFSS[15] is used to review the receiving wire's yield. The fundamental math of the space is a sensible side triangle with various cycles, as found in Fig. 2(a)–(d). Figure 3 shows the virtual return dissatisfaction of the receiving wire (without the tuning opening) for the obvious feature seasons of the Koch estimation, starting with the even triangle. Dazzling The Resonant is a term used to depict a marvel.

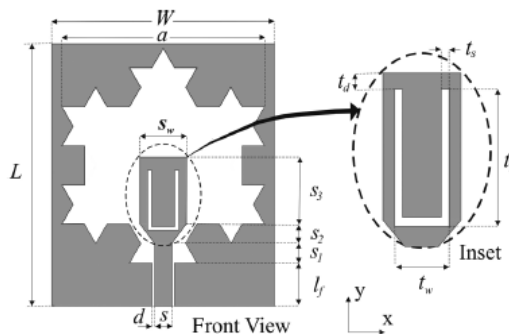


Fig. 1. Proposed slot loaded modified Koch snowflake slot antenna.

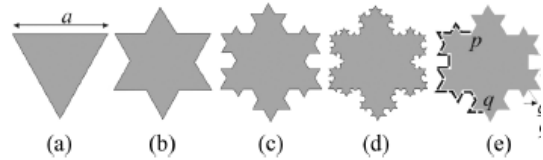


Fig. 2. Koch snowflake geometry in its different iteration stages. (a) Basic geometry. (b) First iteration. (c) Second iteration. (d) Third iteration. (e) Modified second iteration.

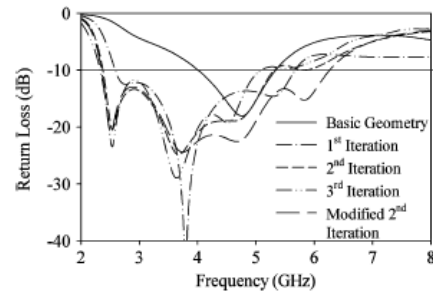


Fig. 3. Simulated return loss of the CPW-fed Koch slot antenna (without the tuning slot)

For The space extent of the fundamental radio wire lessens as the proportion of emphases increases. Ignoring the way that the space edge increases by a factor of 33% for each highlight, the distinction in resonance repeat doesn't follow a relative model. The Koch fractal plan in like manner further makes coupling between the feed stub and the opening, achieving an all the more wide space radio wire impedance information move limit. Right when the proportion of emphases shows up at 2, the receiving wire's functioning band shifts from 4-5.32 GHz to 2.36-5.5 GHz.

The feature development has been improved essentially, allowing the running repeat to drop somewhat. The radio wire's functioning information move limit is arrived at 2.36-6.26 GHz when the condition of the subsequent supplement space is changed fairly as shown in Fig. 2(e), through and through consolidating the WLAN/Wi-MAX parties. HFSS and the electrical field in the fractal opening are used to duplicate the spread of surface stream on the essential layer and the electrical field in the fractal space.

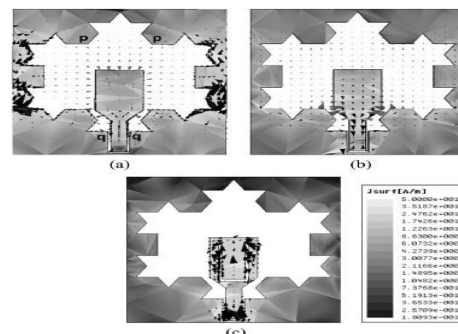


Fig. 4. Simulated current distribution on the patch and the electric field on the slot at (a) 2.5 GHz and (b) 4.55 GHz and (c) 4.55 GHz (with slot).

Those showed in Fig. 4. At the fractal farthest reaches of the opening at the most reduced full repeat, a half recurrent mix in current is found in the spread showed in Fig. 4(a) (2.5 GHz). On this foundation, plan assessments for the proposed radio wire's new turn of events and working repetitive reach are made. The fractal presence of the hidden grants the constraint of the Koch turn not really settled perpetually additionally as the side of the basic equilateral triangle computation. Regardless, when meandered from air as a substrate, the headway of the dielectric establishment further encourages the antennalowering working intermittent tests. Thusly, the rehash at the major resonating is certainly appeared contrastingly relating to the rehash as shown in Fig. 2(e).

$$\frac{\lambda_1}{2} = k \times (pq) \quad (1)$$

where the length of the slot boundary is

$$pq = 17\frac{a}{9} + \frac{a}{3\sqrt{3}} \quad (2)$$

and k is an empirically derived parameter which includes the effect of the substrate. The stub dimensions, shown in Fig. 1, can also be derived in terms of "a" as

$$s_w = \frac{a}{3}, \quad s_1 = s_2 = \frac{a}{6\sqrt{3}}, \quad s_3 = \frac{a}{2\sqrt{3}}. \quad (3)$$

The length limits the size of the ground plane on the feed side to 5 mm, while the ground plane borders on the other three sides are approximately 1 mm away from the slot's vertex.

TABLE I
ANTENNA DETAILS

Antenna	Laminate	ϵ_r	h (mm)	s (mm)	d (mm)	k	a (mm)	L (mm ²)	Computed(mm)				Optimised(mm)			
									f_{r1}	f_{r2}	f_0	h_0	f_{r1}	f_{r2}	f_0	h_0
1	RT/durad® 5880	2.2	1.57	2.8	0.15	1.03	28	35.5x30.5	2.7	8.1	9.33	2.7	10.6	9		
2	Nalco NH9338	3.38	1.57	2.4	0.3	1.07	27	34.5x29.5	2.6	7.8	9	2.6	9.3	8.4		
3	PFA Epoxy	4.4	1.6	2.2	0.3	1.11	26	33.5x28.5	2.5	7.5	8.7	2.5	9.1	6.4		
4	RT/durad® 6006	6.15	1.27	2.0	0.4	1.15	25	32.5x27.5	2.4	7.2	8.3	2.4	7.5	7.2		
5	RT/durad 6010LM	10.2	1.27	2.0	0.7	1.2	24	31.5x26.5	2.3	6.9	8.0	2.3	7.3	5.6		

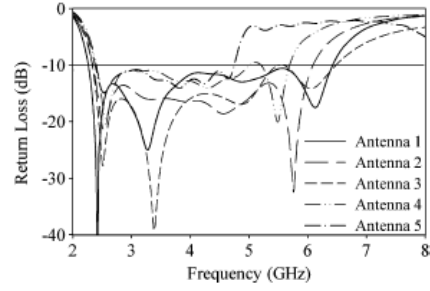


Fig. 5. Return loss of the antennas with parameters as in Table I.

TABLE II
COMPUTED AND MEASURED FOR
DIFFERENT SLOT SIZES OF ANTENNA 3

a (mm)	18	22	26	38
f_{r1} (GHz) Computed	3.6	2.95	2.5	2.32
f_{r1} (GHz) Measured	3.54	2.98	2.51	2.34
Band (GHz)	3.2-7.6	2.74-6.96	2.32-6.5	2.16-5.88

Those showed in Fig. 4. At the fractal furthest ranges of the space at the most diminished shocking repeat, a half recurrent blend in flow is found in the dispersal showed in Fig. 4(a) (2.5 GHz). On this foundation, plan assessments for the proposed radio wire's new turn of events and working intermittent reach are passed on. The fractal presence of the space allows the line of the Koch twist not really settled moreover as the side of the basic equilateral triangle math. Regardless, when veered from air as a substrate, the headway of the dielectric base further encourages the antennalowering working intermittent tests. Thusly, the rehash at the mysterious resonance is precisely stood separated from the rehash as shown in Fig. 2(e)

$$\text{slot length} \approx \frac{c}{2f_{\text{notch}}} \left(\sqrt{\frac{\epsilon_r + 1}{2}} \right)^{-1}. \quad (4)$$

Figure 4(b) and (c) illustrate the surface currents on the patch at 4.5 GHz, with and without the slot. It demonstrates how the antenna's stimulated surface currents interfere in a damaging manner.

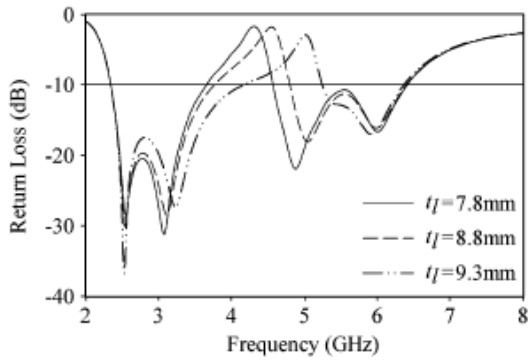


Fig. 6. Return loss of the antenna for different slot lengths

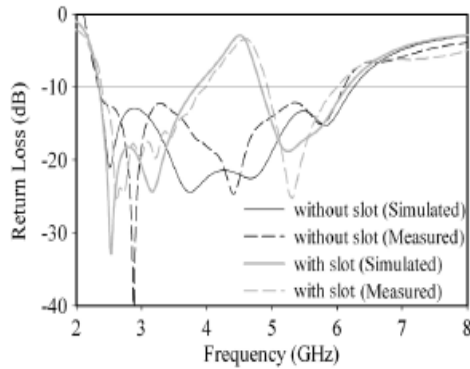


Fig. 7. Return loss of the antenna with and without the slot due to the presence of a slot in the tuning stub mm ,hence causing the antenna to be non-radiating at that frequency.

1. RESULTS

The Rhode andSchwarz ZVB20 Vector Network Analyzer is utilized to deal with the impedance and radiation properties of the proposed recieving wire model (Antenna 3), as displayed in Table I. Figure 7 shows the radio wire's recreated and overviewed bring difficulty back. Simultaneousness with an undeniable degree of execution. The 10 dB move speed of the wide-band recieving wire (without the space) is 3.77 GHz (2.33–6.1 GHz). The radio wire offers twofold wide-band yield in the lower and upper social affairs, with a 10 dB data transmission of 1.57 GHz (2.38-3.95 GHz) and 1.1 GHz, only, by goodness of the tuning opening (4.95-6.05 GHz).

The WLAN bundles 2.4-2.484 GHz, 5.15-5.35 GHz, and 5.725-5.825 GHz, also as 2.5-2.69 GHz, are

consequently

covered.

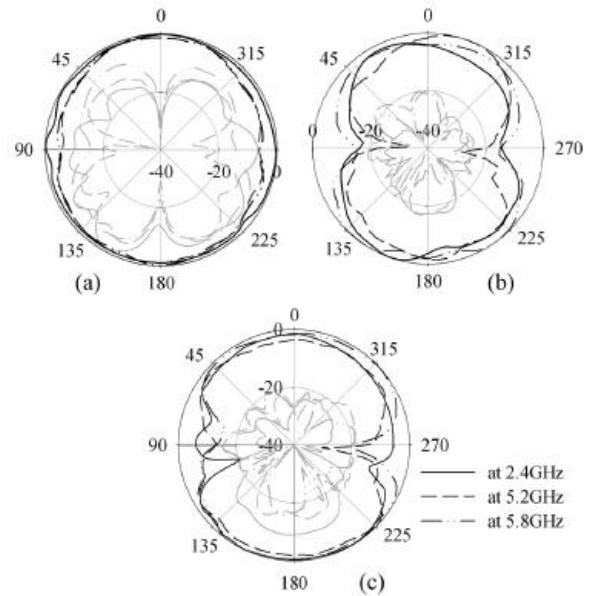


Fig. 8. Radiation pattern of the antenna

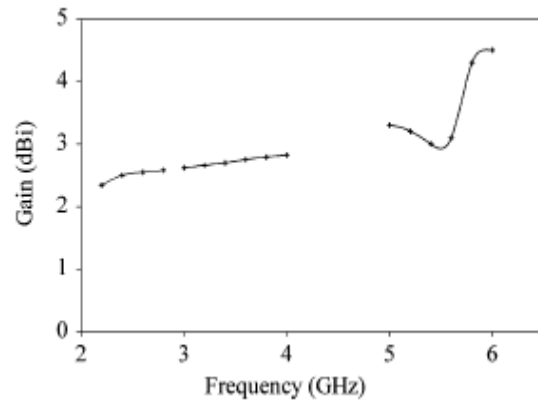


Fig. 9. Gain of the proposed design (Antenna 3).

3.4–3.69 GHz and 5.25–5.85 GHz Wi-MAX bunches Figure 8 shows the radiation plans recorded by the radio wire in the planes. Across the entire working come to, the radiation plans are believed to be predictable, omni-directional, and polarization planes around the turn. Cross-polarization is found in the plane in view of the strong level piece of the electric field, as shown in Fig. 4. (a). Figure 9 shows the expected receiving wire get in the functioning gatherings. In the band generally speaking, the benefit is displayed to remain over 2.0 dBi. The idea of receiving wire radiation in the functioning band is higher than 85%, according to amusement tests.

2. CONCLUSION

The detail for an updated Koch fractal printed CPW-managed slotantenna is given, which is reasonable for WLAN 2.4/5.2/5.8 GHz and Wi-MAX 2.5/3.5/5.5 GHz works out. Reenacted disclosures show that utilizing a Koch fractal space instead of a three-sided opening shape reduces the rehash of association with wide-band arranging. The radio wire is near nothing, even with the land plane, and a quick change opening guarantees twofold wide-band advancement over WLAN and WiMAX frequencies. Intelligent still open to question and checked to address the radio wire on various substrates. The course of action is fitting for far off broadband systems association applications considering the way that to its gigantic impedance data move limit and truly predictable and omnidirectional radiation plans.

REFERENCES

- [1] M. Kahrizi, T. K. Sarkar, and Z. A. Maricevic, "Analysis of a wideradiating slot in the ground plane of a microstrip line," *IEEE Trans. Microw. Theory Tech.*, vol. 41, no. 1, pp. 29–37, Jan. 1993.
- [2] J.-Y. Chiou, J.-Y. Sze, and K. L. Wong, "A broadband CPW-Fed striploaded square slot antenna," *IEEE Trans. Antennas Propag.*, vol. 51, no. 4, pp. 719–721, Apr. 2003.
- [3] H.-D. Chen, "Broadband CPW-Fed square slot antennas with awidened tuning stub," *IEEE Trans. Antennas Propag.*, vol. 51, no. 4, pp. 1982–1986, Aug. 2003.
- [4] J.-Y. Jan and C.-Y. Hsiang, "Wideband CPW-fed slot antenna for DCS, PCS, 3 G and Bluetooth bands," *Electron. Lett.*, vol. 42, no. 24, pp. 1377–1378, Nov. 2006.
- [5] C.-J. Wang and J.-J. Lee, "A pattern-frequency-dependent wide-band slot antenna," *IEEE Antennas Wireless Propag. Lett.*, vol. 5, pp. 65–68, 2006.
- [6] S.-W. Qu, C. Ruan, and B.-Z. Wang, "Bandwidth enhancement of wide-slot antenna fed by CPW and microstrip line," *IEEE Antennas Wireless Propag. Lett.*, vol. 5, pp. 15–17, 2006.
- [7] E. S. Angelopoulos, A. Z. Anastopoulos, D. I. Kaklamani, A. A. Alexandridis, F. Lazarakis, and K. Dangakis, "Circular

and elliptical CPW-Fed slot and microstrip-fed antennas for ultrawide-band applications," *IEEE Antennas Wireless Propag. Lett.*, vol. 5, pp. 294–297, 2006.

[8] Y.-C. Lin and K.-J. Hung, "Compact ultrawide-band rectangular aperture antenna and band-Notched designs," *IEEE Trans. Antennas Propag.*, vol. 54, no. 11, pp. 3075–3081, Nov. 2006.

[9] P. Li, J. Liang, and X. Chen, "Study of printed elliptical/circular slot antennas for ultrawide-band applications," *IEEE Trans. Antennas Propag.*, vol. 54, no. 6, pp. 1670–1675, Jun. 2006.

[10] W.-S. Chen, "A novel broadband design of a printed rectangular slot antenna for wireless applications," *Microw. J.*, vol. 49, no. 1, pp. 122–130, 2006.

[11] W.-S. Chen and F.-M. Hsieh, "A broadband design for a printed isosceles triangular slot antenna for wireless communications," *Microw. J.*, vol. 48, no. 7, pp. 98–112, 2005.

[12] D. H. Werner and S. Ganguly, "An overview of fractal antenna engineering research," *IEEE Antennas Propag. Mag.*, vol. 45, no. 1, pp. 38–57, Feb. 2003.

[13] W.-S. Lee, W.-G. Lim, and J.-W. Yu, "Multiple band Notched planar monopole antenna for multiband wireless systems," *IEEE Microw. Wireless Comp. Lett.*, vol. 15, no. 9, pp. 576–578, Sep. 2005.

[14] T. Dissanayake and K. Esselle, "Prediction of Notched frequency of slot-loaded printed slot antennas," *IEEE Trans. Antennas Propag.*, vol. 55, no. 11, pp. 3320–3325, Nov. 2007.

[15] Ansoft HFSS v. 9.0: Ansoft Inc. Pittsburgh, PA.